## CHAPTER 4

## APPLICATIONS OF BOOLEAN ALGEBRA MINTERM AND MAXTERM EXPANSIONS



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## Conversion of English Sentences to Boolean Equations

The three main steps in designing a single-output combinational switching circuit are

1. Find a switching function that specifies the desired behavior of the circuit.
2. Find a simplified algebraic expression for the function.
3. Realize the simplified function using available logic elements.

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## Example 1

## Mary watches TV if it is Monday night and she has finished her homework.

F
A
B

We will define a two-valued variable to indicate the truth of falsity of each phrase:
$F=1$ if "Mary watches TV" is true; otherwise $F=0$.
$A=1$ if "it is Monday night" is true; otherwise $A=0$.
$B=1$ if "she has finished her homework" is true;
otherwise $B=0$.

Because $F$ is "true" if $A$ and $B$ are both "true", we can represent the sentence by $F=A \cdot B$

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## Example 2

The alarm will ring iff the alarm switch is turned on and the door is not closed, or it is after 6 P.M. and the window is not closed.


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## Example 2 (continued)

The alarm will ring iff the alarm switch is turned on and the door is not closed, or it is after 6 P.M. and the window is not closed.

The corresponding equation is:

$$
Z=A B^{\prime}+C D^{\prime}
$$

And the corresponding circuit is:


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## Combinational Logic Design using a Truth Table

Suppose we want the output of a circuit to be $f=1$ if $\mathrm{N} \geq 011_{2}$ and $f=0$ if $\mathrm{N}<011_{2}$.
Then the truth table is:

| $A$ | $B$ | $C$ | $f$ | $f^{\prime}$ |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 1 |
| 0 | 0 | 1 | 0 | 1 |
| 0 | 1 | 0 |  | 0 |
| 0 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 1 | 0 |
| 1 | 0 | 1 | 1 | 0 |
| 1 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 1 | 0 |



Figure 4-1
(b)

Next, we will derive an algebraic expression for $f$ from the truth table by using the combinations of values of $A, B$, and $C$ for which $f=1$. For example, the term $\mathrm{A}^{\prime} \mathrm{BC}$ is 1 only if $\mathrm{A}=0$, $B=1$, and $C=1$. Finding all terms such that $f=1$ and ORing them together yields:
$f=A^{\prime} B C+A B^{\prime} C^{\prime}+A B^{\prime} C+A B C^{\prime}+A B C$

The equation can be simplified by first combining terms and then eliminating $A^{\prime}$ :

$$
\begin{equation*}
f=A^{\prime} B C+A B^{\prime}+A B=A^{\prime} B C+A=A+B C \tag{4-2}
\end{equation*}
$$

This equation leads directly to the following circuit:


Instead of writing $f$ in terms of the 1 's of the function, we may also write $f$ in terms of the 0 's of the function. Observe that the term $\mathrm{A}+\mathrm{B}+\mathrm{C}$ is 0 only if $\mathrm{A}=\mathrm{B}=\mathrm{C}=0$. ANDing all of these ' 0 ' terms together yields:

$$
\begin{equation*}
f=(A+B+C)\left(A+B+C^{\prime}\right)\left(A+B^{\prime}+C\right) \tag{4-3}
\end{equation*}
$$

By combining terms and using the second distributive law, we can simplify the equation:

$$
\begin{align*}
& f=(A+B+C)\left(A+B+C^{\prime}\right)\left(A+B^{\prime}+C\right)  \tag{4-3}\\
& f=(A+B)\left(A+B^{\prime}+C\right)=A+B\left(B^{\prime}+C\right)=A+B C \tag{4-4}
\end{align*}
$$

## Minterm and Maxterm Expansions

$$
f=A^{\prime} B C+A B^{\prime} C^{\prime}+A B^{\prime} C+A B C^{\prime}+A B C(4-1)
$$

Each of the terms in Equation (4-1) is referred to as a minterm. In general, a minterm of $n$ variables is a product of $n$ literals in which each variable appears exactly once in either true or complemented form, but not both.
(A literal is a variable or its complement)

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## Table 4-1 Minterms and Maxterms for Three Variables

| Row No. | $A B C$ | Minterms | Maxterms |
| :---: | :---: | :---: | :---: |
| 0 | 000 | $A^{\prime} B^{\prime} C^{\prime}=m_{0}$ | $A+B+C=M_{0}$ |
| 1 | 001 | $A^{\prime} B^{\prime} C=m_{1}$ | $A+B+C^{\prime}=M_{1}$ |
| 2 | 010 | $A^{\prime} B C^{\prime}=m_{2}$ | $A+B^{\prime}+C=M_{2}$ |
| 3 | 011 | $A^{\prime} B C=m_{3}$ | $A+B^{\prime}+C^{\prime}=M_{3}$ |
| 4 | 100 | $A B^{\prime} C^{\prime}=m_{4}$ | $A^{\prime}+B+C=M_{4}$ |
| 5 | 101 | $A B^{\prime} C=m_{5}$ | $A^{\prime}+B+C^{\prime}=M_{5}$ |
| 6 | 110 | $A B C^{\prime}=m_{6}$ | $A^{\prime}+B^{\prime}+C=M_{6}$ |
| 7 | 111 | $A B C=m_{7}$ | $A^{\prime}+B^{\prime}+C^{\prime}=M_{7}$ |

Minterm expansion for a function is unique. Equation (4-1) can be rewritten in terms of $m$-notation as:
$f=A^{\prime} B C+A B^{\prime} C^{\prime}+A B^{\prime} C+A B C^{\prime}+A B C(4-1)$
$f(A, B, C)=m_{3}+m_{4}+m_{5}+m_{6}+m_{7}$

This can be further abbreviated by listing only the decimal subscripts in the form:
$f(A, B, C)=\Sigma m(3,4,5,6,7)$

## Minterm Expansion Example

Find the minterm expansion of $f(a, b, c, d)=a^{\prime}\left(b^{\prime}+d\right)+a c d^{\prime}$.

$$
\begin{align*}
& f=a^{\prime} b^{\prime}+a^{\prime} d+a c d^{\prime} \\
& f=a^{\prime} b^{\prime}+a^{\prime} d+a c d^{\prime} \\
& =a^{\prime} b^{\prime}\left(c+c^{\prime}\right)\left(d+d^{\prime}\right)+a^{\prime} d\left(b+b^{\prime}\right)\left(c+c^{\prime}\right)+\operatorname{acd}\left(b+b^{\prime}\right) \\
& =a^{\prime} b^{\prime} c^{\prime} d^{\prime}+a^{\prime} b^{\prime} c^{\prime} d+a^{\prime} b^{\prime} c d^{\prime}+a^{\prime} b^{\prime} c d+a^{\prime} b^{\prime} c^{\prime} d+a^{\prime} b^{\prime} c d \\
& +a^{\prime} b c^{\prime} d+a^{\prime} b c d+a b c d^{\prime}+a b^{\prime} c d^{\prime} \\
& \text { (4-9) } a b c d^{\prime}+a b^{\prime} c d^{\prime} \\
& f=a^{\prime} b^{\prime} c^{\prime} d^{\prime}+a^{\prime} b^{\prime} c^{\prime} d+a^{\prime} b^{\prime} c d^{\prime}+a^{\prime} b^{\prime} c d+a^{\prime} b c^{\prime} d+a^{\prime} b c d+a b c d^{\prime}+a b^{\prime} c d^{\prime} \\
& 0000 \quad 0001 \quad 0010 \quad 0011 \quad 0101 \quad 0111 \quad 1110 \quad 1010 \\
& f=\Sigma m(0,1,2,3,5,7,10,14) \tag{4-10}
\end{align*}
$$

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## Maxterm Expansion Example

Find the maxterm expansion of $f(a, b, c, d)=a^{\prime}\left(b^{\prime}+d\right)+a c d^{\prime}$.

$$
\begin{align*}
& f=a^{\prime}\left(b^{\prime}+d\right)+a c d^{\prime} \\
& =\left(a^{\prime}+c d^{\prime}\right)\left(a+b^{\prime}+d\right)=\left(a^{\prime}+c\right)\left(a^{\prime}+d^{\prime}\right)\left(a+b^{\prime}+d\right) \\
& =\left(a^{\prime}+b b^{\prime}+c+d d^{\prime}\right)\left(a^{\prime}+b b^{\prime}+c c^{\prime}+d^{\prime}\right)\left(a+b^{\prime}+c c^{\prime}+d\right) \\
& =\left(a^{\prime}+b b^{\prime}+c+d\right)\left(a^{\prime}+b b^{\prime}+c+d^{\prime}\right)\left(a^{\prime}+b b^{\prime}+c+d^{\prime}\right) \\
& \left(a^{\prime}+b b^{\prime}+c^{\prime}+d^{\prime}\right)\left(a+b^{\prime}+c c^{\prime}+d\right) \\
& \begin{array}{ccc}
\left(a^{\prime}+b+c+d\right)\left(a^{\prime}+b^{\prime}+c+d\right)\left(a^{\prime}+b+c+d^{\prime}\right)\left(a^{\prime}+b^{\prime}+c+d^{\prime}\right) \\
1000 & 1100 & 1001
\end{array} \\
& \begin{array}{ccc}
\left(a^{\prime}+b+c^{\prime}+d^{\prime}\right)\left(a^{\prime}+b^{\prime}+c^{\prime}+d^{\prime}\right)\left(a+b^{\prime}+c+d\right)\left(a+b^{\prime}+c^{\prime}+d\right) \\
1011 & 01111 & 0110
\end{array} \\
& =\Pi M(4,6,8,9,11,12,13,15) \tag{4-11}
\end{align*}
$$

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Table 4-2. General Truth Table for Three Variables

Table 4-2 represents a truth table for a general function of three
variables. Each $\mathrm{a}_{\mathrm{i}}$ is a constant with a value of 0 or 1.

| $A$ | $B$ | $C$ | $F$ |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | $a_{0}$ |
| 0 | 0 | 1 | $a_{1}$ |
| 0 | 1 | 0 | $a_{2}$ |
| 0 | 1 | 1 | $a_{3}$ |
| 1 | 0 | 0 | $a_{4}$ |
| 1 | 0 | 1 | $a_{5}$ |
| 1 | 1 | 0 | $a_{6}$ |
| 1 | 1 | 1 | $a_{7}$ |

$$
F=a_{0} m_{0}+a_{1} m_{1}+a_{2} m_{2}+\cdots+a_{7} m_{7}=\sum_{i=0}^{7} a_{i} m_{i}
$$

## General Minterm and Maxterm Expansions

We can write the minterm expansion for a general function of three variables as follows:
$F=a_{0} m_{0}+a_{1} m_{1}+a_{2} m_{2}+\cdots+a_{7} m_{7}=\sum_{i=0}^{7} a_{i} m_{i}$
The maxterm expansion for a general function of three variables is:
$F=\left(a_{0}+M_{0}\right)\left(a_{1}+M_{1}\right)\left(a_{2}+M_{2}\right) \cdots\left(a_{7}+M_{7}\right)=\prod_{i=0}^{7}\left(a_{i}+M_{i}\right)$

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## Table 4-3 summarizes the procedures for conversion between minterm and maxterm expansions of $F$ and $F^{\prime}$

## Table 4-3. Conversion of Forms

|  |  | DESIRED FORM |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Minterm Expansion of $F$ | Maxterm Expansion of $F$ | Minterm Expansion of $F^{\prime}$ | Maxterm Expansion of $F^{\prime}$ |
| $\sum_{\substack{0}}^{\substack{0}}$ | Minterm Expansion of $F$ |  | maxterm nos. are those nos. not on the minterm list for $F$ | list minterms not present in $F$ | maxterm nos. are the same as minterm nos. of $F$ |
| $\underset{\sim}{\mathbf{u}}$ | Maxterm Expansion of $F$ | minterm nos. are those nos. not on the maxterm list for $F$ | - | minterm nos. are the same as maxterm nos. of $F$ | list maxterms not present in $F$ |

## Table 4-4. Application of Table 4-3

|  | DESIRED FORM |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | Minterm Expansion of $f$ | Maxterm Expansion of $f$ | Minterm Expansion of $f^{\prime}$ | Maxterm Expansion of $f^{\prime}$ |
| $\begin{aligned} & \sum_{\mathrm{O}} \overline{f=} \\ & \text { O } \Sigma m(3,4,5,6,7) \end{aligned}$ |  | $\Pi M(0,1,2)$ | $\Sigma m(0,1,2)$ | $\Pi M(3,4,5,6,7)$ |
| $\begin{aligned} & \underset{\Psi}{\underset{\sim}{u}} f= \\ & \Pi M(0,1,2) \end{aligned}$ | $\Sigma m(3,4,5,6,7)$ |  | $\Sigma m(0,1,2)$ | $П M(3,4,5,6,7)$ |

## Incompletely Specified Functions

A large digital system is usually divided into many subcircuits. Consider the following example in which the output of circuit $\mathrm{N}_{1}$ drives the input of circuit $\mathrm{N}_{2}$ :


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Let us assume the output of $\mathrm{N}_{1}$ does not generate all possible combinations of values for $A, B$, and $C$. In particular, we will assume there are no combinations of values for $w, x, y$, and $z$ which cause $A, B$, and $C$ to assume values of 001 or 110 .

Table 4-5: Truth Table with Don't Cares

| $A$ | $B$ | $C$ | $F$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 0 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | $X$ |
| 1 | 1 | 1 | 1 |

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When we realize the function, we must specify values for the don't-cares. It is desirable to choose values which will help simplify the function. If we assign the value 0 to both X 's, then

$$
F=A^{\prime} B^{\prime} C^{\prime}+A^{\prime} B C+A B C=A^{\prime} B^{\prime} C^{\prime}+B C
$$

If we assign 1 to the first X and 0 to the second, then

$$
F=A^{\prime} B^{\prime} C^{\prime}+A^{\prime} B^{\prime} C+A^{\prime} B C+A B C=A^{\prime} B^{\prime}+B C
$$

If we assign 1 to both $X^{\prime}$ s, then

$$
\begin{aligned}
F & =A^{\prime} B^{\prime} C^{\prime}+A^{\prime} B^{\prime} C+A^{\prime} B C+A B C^{\prime}+A B C \\
& =A^{\prime} B^{\prime}+B C+A B
\end{aligned}
$$

The second choice of values leads to the simplest solution.

The minterm expansion for Table $4-5$ is:

$$
F=\Sigma m(0,3,7)+\Sigma d(1,6)
$$

The maxterm expansion for Table $4-5$ is:
$F=\Pi M(2,4,5) \bullet \Pi D(1,6)$

Table 4-5

| $A$ | $B$ | $C$ | $F$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 |
| 0 | 0 | 1 | $X$ |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | $X$ |
| 1 | 1 | 1 | 1 |

## Examples of Truth Table Construction

We will design a simple binary adder that adds two 1-bit binary numbers, $a$ and $b$, to give a 2-bit sum. The numeric values for the adder inputs and outputs are as follows:


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We will represent inputs to the adder by the logic variables $A$ and $B$ and the 2-bit sum by the logic variables $X$ and $Y$, and we construct a truth table:

| $A$ | $B$ | $X$ | $Y$ |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 |
| 0 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 |
| 1 | 1 | 1 | 0 |

Because a numeric value of 0 is represented by a logic 0 and a numeric value of 1 by a logic 1 , the 0 's and 1 's in the truth table are exactly the same as in the previous table. From the truth table,

$$
X=A B \text { and } Y=A^{\prime} B+A B^{\prime}=A \oplus B
$$

Ex: Design an adder which adds two 2-bit binary numbers to give a 3-bit binary sum. Find the truth table for the circuit. The circuit has four inputs and three outputs as shown:


TRUTH TABLE:

| $N_{1}$ | $\mathrm{N}_{2}$ | $\mathrm{N}_{3}$ | $N_{1}$ | $\mathrm{N}_{2}$ | $\mathrm{N}_{3}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $A B$ | $C D$ | $X Y Z$ | A B | $C D$ | $X Y Z$ |
| 00 | 00 | 000 | 10 | 00 | 010 |
| 00 | 01 | 001 | 10 | 01 | 011 |
| 00 | 10 | 010 | 10 | 10 | 100 |
| 00 | 11 | 011 | 10 | 11 | 101 |
| 01 | 00 | 001 | 11 | 00 | 011 |
| 01 | 01 | 010 | 11 | 01 | 100 |
| 01 | 10 | 011 | 11 | 10 | 101 |
| 01 | 11 | 100 | 11 | 11 | 110 |

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## Design of Binary Adders and Subtractors

We will design a parallel adder that adds two 4-bit unsigned binary numbers and a carry input to give a 4-bit sum and a carry output.

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Figure 4-2: Parallel Adder for 4-Bit Binary Numbers

One approach would be to construct a truth table with nine inputs and five outputs and then derive and simplify the five output equations.

A better method is to design a logic module that adds two bits and a carry, and then connect four of these modules together to form a 4-bit adder.

end-around carry for 1's complement

Figure 4-3: Parallel Adder Composed of Four Full Adders


Figure 4-4: Truth Table for a Full Adder

## Full Adder Logic Equations

The logic equations for the full adder derived from the truth table are:

$$
\begin{align*}
\text { Sum } & =X^{\prime} Y^{\prime} C_{\text {in }}+X^{\prime} Y C_{\text {in }}^{\prime}+X Y^{\prime} C_{\text {in }}^{\prime}+X Y C_{\text {in }} \\
& =X^{\prime}\left(Y^{\prime} C_{\text {in }}+Y C_{\text {in }}^{\prime}\right)+X\left(Y^{\prime} C_{\text {in }}^{\prime}+Y C_{\text {in }}\right)  \tag{4-20}\\
& =X^{\prime}\left(Y \oplus C_{\text {in }}\right)+X\left(Y \oplus C_{\text {in }}\right)^{\prime}=X \oplus Y \oplus C_{\text {in }} \\
C_{\text {out }} & =X^{\prime} Y C_{\text {in }}+X Y^{\prime} C_{\text {in }}+X Y C_{\text {in }}^{\prime}+X Y C_{\text {in }} \\
& =\left(X^{\prime} Y C_{\text {in }}+X Y C_{\text {in }}\right)+\left(X Y^{\prime} C_{\text {in }}+X Y C_{\text {in }}\right)+\left(X Y C_{\text {in }}^{\prime}+X Y C_{\text {in }}\right)  \tag{4-21}\\
& =Y C_{\text {in }}+X C_{\text {in }}+X Y
\end{align*}
$$

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## cosis Sum



Figure 4-5: Implementation of Full Adder

## Overflow for Signed Binary Numbers

An overflow has occurred if adding two positive numbers gives a negative result or adding two negative numbers gives a positive result.

We define an overflow signal, $\mathrm{V}=1$ if an overflow occurs. For Figure 4-3, $V=A_{3}{ }^{\prime} B_{3}{ }^{\prime} S_{3}+A_{3} B_{3} S_{3}{ }^{\prime}$

end-around carry for 1 's complement
Figure 4-3

Full Adders may be used to form A - B using the 2's complement representation for negative numbers.
The 2's complement of B can be formed by first finding the 1 's complement and then adding 1.


Figure 4-6: Binary Subtracter Using Full Adders

Alternatively, direct subtraction can be accomplished by employing a full subtracter in a manner analogous to a full adder.


Figure 4-7: Parallel Subtracter

## Table 4.6. Truth Table for Binary Full Subtracter

| $x_{i}$ | $y_{i}$ | $b_{i}$ | $b_{i+1}$ |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 |  |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 | 1 |
| 0 | 1 | 1 | 1 |
| 1 | 1 | 1 |  |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 0 |
| 1 | 0 | 0 | 0 |
| 1 | 1 | 1 |  |

## Consider $x_{i}=0, y_{i}=1$, and $b_{i}=1$ :

|  | Column $i$ <br> Before <br> Borrow | Column $i$ <br> After <br> Borrow |
| ---: | :---: | :---: |
| $x_{i}$ | 0 | 10 |
| $-b_{i}$ | -1 | -1 |
| $\frac{-y_{i}}{d_{i}}$ | -1 | $\frac{-1}{0} \quad\left(b_{i+1}=1\right)$ |

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